## DLDA

## (CBCGS MAY 2018)

Q1] a) Write the entity declaration in VHDL for NOR gate.

## Solution :-

NOR Gate:
A logic gate whose output logic is ' 1 ' if and only if all of its inputs are logic ' 0 '.
Entity declaration of NOR gate
Library IEEE;
use IEEE.STD_LOGIC_1164.ALL;
entity nor_df is
Port (a: in STD_LOGIC;
b: in STD_LOGIC;
c: out STD_LOGIC);
end nor_df;
architecture Behavioral of nand_df
is
begin
c<=a nor b;

Q1] b) Add (22) ${ }_{10}$ to (56) ${ }_{10}$ in BCD.

## Solution :-

BCD of $(22)_{10}$ is 00100010
BCD of $(56)_{10}$ is 01010110

00100010

+ 01010110

01111000
$(22)_{10}+(56)_{10}=01111000$

## Q1] c) Convert decimal 57 into binary, base 7 and hexadecimal.

## Solution :-

Decimal 57 to binary

| 2 | 57 |  |
| :--- | :--- | :--- |
| 2 | 28 | 1 |
| 2 | 14 | 0 |
| 2 | 7 | 0 |
| 2 | 3 | 1 |
| 2 | 1 | 1 |
|  | 0 | 1 |

Decimal 57 to base 7

| 7 | 57 |  |
| :---: | :---: | :---: |
| 7 | 8 |  |
| 7 | 1 | 1 |
|  | 0 | 1 |

Decimal 57 to hexadecimal

| 16 | 57 |  |
| :--- | :--- | :--- |
| 16 | 3 | 9 |
|  | 0 | 3 |

Q1] d) Construct hamming code for 1010.

## Solution :-

7 bit Hamming code is given as

| D7 | D6 | D5 | P4 | D3 | P2 | P1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Given 4 bit is 1010 therefore it can be written as

| 1 | 0 | 1 | P4 | 0 | P2 | P1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

P1: For P1 consider P1, D3, D5, D7
D3, D5, D7 = 011
For even parity set P1 = 0

P1: For P2 consider P2, D3, D6, D7
D3, D6, D7 = 001
For even parity set P2 = 1
P1: For P1 consider P4, D5, D6, D7
D5, D6, D7 = 011
For even parity set P4 $=0$

| 1 | 0 | 1 | 0 | 0 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Therefore, hamming code for 1010 is given as 1010010.

Q1] e) Perform subtraction using 2's complement for (10) 10 $_{10}$ ( 7$)_{10}$.

## Solution :-

Binary representation of 7 is 0111
1 's complement of 7 is 1000
2 's complement of 7 is 1 's complement +1
1000
$+\quad 1$

1001
Now adding 10 and 2 's complement of 7
1010

+ 1001

10011

Discard carry 1
Therefore $(\mathbf{1 0})_{10}-(\mathbf{7})_{10}=0011=(3)_{10}$.

Q1] f) State and prove De Morgan's law.

## Solution :-

Theorem 1:

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Statement: The complement of the product of two or more variables is equal to the sum of the components of the variables.
Logical expression: $\overline{\mathrm{A} . \mathrm{B}}=\overline{\mathrm{A}}+\overline{\mathrm{B}}$
NAND is equal to BUBBLED OR.

| A | B | A.B | $\overline{\text { A. } \bar{x}}$ | $\overline{\mathrm{~A}}$ | $\overline{\mathrm{~B}}$ | $\overline{\mathrm{~A}}+\overline{\mathrm{B}}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 0 | 0 | 0 | 0 |

From the above truth table we can see that $\overline{\mathrm{A} . \mathrm{B}}=\overline{\mathrm{A}}+\overline{\mathrm{B}}$ Hence proved.
Theorem 2 :-
Statement: The complement of the sum of two or more variables is equal to the product of the complement of the variables
Logical expression: $\overline{A+B}=\bar{A} . \bar{B}$
NOR is equal to BUBBLED AND.

| $A$ | $B$ | $\bar{A}+\mathrm{B}$ | $\overline{\mathrm{A}+\mathrm{B}}$ | $\overline{\mathrm{A}}$ | $\overline{\mathrm{B}}$ | $\overline{\mathrm{A}} \cdot \overline{\mathrm{B}}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| 0 | 1 | 1 | 0 | 1 | 0 | 0 |
| 1 | 0 | 1 | 0 | 0 | 1 | 0 |
| 1 | 1 | 1 | 0 | 0 | 0 | 0 |

From the above truth table we can see that $\overline{A+B}=\bar{A} \cdot \bar{B}$ Hence proved.

## Q1] g) Convert (77) ${ }_{10}$ into Excess 3 code.

## Solution :-

Excess 3 code is 3 added to BCD.
BCD of 77 is 01110111 and $0 f 3$ is 0011

## Q1] h) Perform addition of (34) $)_{8}$ and (62)8.

(02)

## Solution :-

$(34)_{8}+(62)_{8}$
34
$+62$

116
In the above sum $6+3$ is greater than 7 hence can't be represented as octal therefore we need to divide it with 8


Remainder is the sum and the quotient is carry.
Therefore, we get $(34)_{8}+(62)_{8}=(116)_{8}$.

Q1] i) Find ${ }^{8}$ 's complement of the number (37) 8 and $(301)_{8}$.

## Solution :-

To find 8's complement we first need to find 7's complement and then add 1 to it.
7 's complement of 37 is obtain by subtracting 37 from 77
i) 77
$+\quad 37$

7's complement
$+\quad 1$

41
8's complement
ii) $\quad 777$

```
    + 301
```

476
7's complement
$+\quad 1$

477
8's complement

## Q1] j) Explain ASCII code in brief.

## Solution :-

ASCII was developed by the American National Standards Institute (ANSI). Pronounced ask-ee, ASCII is the acronym for the American Standard Code for Information Interchange.
It is a code for representing 128 English characters as numbers.
Technically, ASCII is 7 -bit representing only 128 characters ( $0-127$ ). The range $0-31$ are control characters, with 32-127 representing alphabetical characters from $A$ to $Z$, numerals from 0 to 9 and punctuation marks (though not in that order).
For example, the ASCII code for uppercase M is 77.
Most computers use ASCII codes to represent text, which makes it possible to transfer data from one computer to another.
Text files stored in ASCII format are sometimes called ASCII files. Text editors and word processors are usually capable of storing data in ASCII format, although ASCII format is not always the default storage format.

Q2] a) Simplify the following equation using $K$ map to obtain SOP equation and realize the minimum equation using only NAND gates.
$F(A, B, C, D)=\Sigma m(1,2,4,6,9,10,12,14)+d(3,7,13)$

## Solution :-

Step 1: Draw K map

|  | 00 | 01 | 11 |  |
| :---: | :---: | :---: | :---: | :---: |

Step 2: SOP Equation
$\mathrm{Y}=\mathrm{B} \bar{C} \bar{D}+\bar{B} \bar{C} D+\mathrm{C} \bar{D}$
Step 3: NAND gate realization
Take double inversion
$Y=\overline{\overline{\mathrm{B}} \overline{\bar{C}} \bar{D}+\bar{B} \bar{C} D+\mathrm{C} \overline{\bar{D}}}$
$\mathrm{Y}=\overline{\bar{B} \bar{C} \bar{D}} \overline{\bar{B}} \bar{C} D \bar{C} \overline{\bar{D}}$


Q2] b) Implement full adder using 8:1 mux.
(10)

## Solution :-

Truth table of full adder

| A | B | Cin | S | Cout |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 1 | 0 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 0 | 1 |
| 1 | 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 1 | 1 |

In full adder there are 3 input $\mathrm{A}, \mathrm{B}$ and Cin and they are added and generates two output S and Cout. $S$ is the sum of $A, B$ and Cin and Cout is the carry generated by adding 3 input's.

## Full Adder Implementation using 8:1 MUX

An 8:1 multiplexer consists of eight data inputs 10 to 17 , three input select lines $A, B$ and $C$ and a single output line $Y$. Depending on the select lines combinations, multiplexer decodes the inputs.
Since the number of input data bits given to the MUX are eight therefore 3 bits (23=8) are needed to select one of the eight data bits. That is there is 3 select lines in $8: 1$ MUX.
Since in full adder there are 2 output, we need two $8: 1$ multiplexer. Output of $1^{\text {st }}$ multiplexer is sum of adder and output of $2^{\text {nd }}$ multiplexer is carry generated of full adder.


Fig. Full Adder using 8:1 MUX
As we can see in the above diagram there are 3 select lines $A, B$ and $C$. there are 7 input 10 to $I 7$. The input is connected to 0 or 1 depending upon the value given in truth table.
The select lines of both multiplexer is connected to same input.

Q3] a) Obtain the minimal expression using Quine Mc-Cluskey method
$F(A, B, C, D)=\Sigma m(1,2,3,5,6,10.11,13,14)+d(4,7)$

## Solution :-

Table 1

| Group | Minterm | Binary Representation |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | A | B | C | D |
|  | M1 | 0 | 0 | 0 | 1 |
|  | M2 | 0 | 0 | 1 | 0 |
|  | M4 $^{*}$ | 0 | 1 | 0 | 0 |
| 1 | m3 | 0 | 0 | 1 | 1 |
|  | m5 | 0 | 1 | 0 | 1 |


|  | m6 | 0 | 1 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  | $m 10$ | 1 | 0 | 1 | 0 |
|  | $m 13$ | 1 | 1 | 0 | 1 |
|  | m14 | 1 | 1 | 1 | 0 |
|  | m7* | 0 | 1 | 1 | 1 |

Table 2

| Group | Minterm | Binary Representation |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | A | B | C | D |
| 0 | m1-m3 | 0 | 0 |  | 1 |
|  | m1-m5 | 0 |  | 0 | 1 |
|  | m2-m3 | 0 | 0 | 1 | - |
|  | m2-m6 | 0 |  | 1 | 0 |
|  | m2-m10 | - | 0 | 1 | 0 |
|  | m4-m5* | 0 | 1 | 0 | - |
|  | m4-m6* | 0 | 1 | - | 0 |
| 1 | m3-m7* |  | 0 |  | 1 |
|  | M5-m13 |  | 1 | 0 | 1 |
|  | m5-m7* | 0 | 1 |  | 1 |
|  | m6-m14 |  | 1 | 1 | 0 |
|  | m6-m7* | 0 | 1 | 1 |  |
|  | m10-m14 | 1 |  | 1 | 0 |

Table 3

| Group | Minterm | Binary Representation |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | A | B | C | D |
| 0 | m1-m3-m5-m7 | 0 | - | - | 1 |
|  | M1-m5-m3-m7 | 0 | - | - | 1 |
|  | M1-m5-m10-m14 | - | - |  | - |
|  | M2-m3-m6-m7 | 0 | - | 1 | - |
|  | M2-m6-m3-m7 | 0 | - | 1 | , |
|  | M2-m6-m10-m14 | - | - | 1 | 0 |
|  | $\mathrm{M} 2-\mathrm{m} 10-\mathrm{m} 5=\mathrm{m} 13$ | - | - |  |  |
|  | M2-m10-m6-m14 |  | _ | 1 | 0 |
|  | M4-m5-m6-m7 | 0 | 1 | - | - |
|  | M4-m6-m5-m7 | 0 | 1 | - | - |


| PI | Decimal of PI | 1 | 2 | 3 | 5 | 6 | 10 | 13 | 14 | 4 | 7 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\bar{A} D$ | M 1 m 3 m 5 m 7 | $\otimes$ |  | X | X |  |  |  |  |  | X |
| $\bar{A} C$ | M 2 m 3 m 6 m 7 |  | X | X |  | X |  |  |  |  | X |
| $C \bar{D}$ | $\mathrm{M} 2-\mathrm{m} 6-\mathrm{m} 10-\mathrm{m} 14$ |  | X |  |  | X | $\otimes$ |  | $\otimes$ |  |  |
| $\bar{A} B$ | $\mathrm{M} 4-\mathrm{m} 5-\mathrm{m} 6=\mathrm{m} 7$ |  |  |  | X | X |  |  |  | $\otimes$ | X |

$$
Y=\bar{A} \mathbf{D}+\overline{\mathbf{C D}}+\overline{\mathbf{A}} \mathbf{B}
$$

## Q3] b) What is race around condition? How to overcome it?

## Solution:-

The race around condition occurs in JK flipflop when both the inputs are high that is J=K=1 normally in level triggered JK flipflop.


Fig.3.1 Race around condition
The above figure shows the race around condition in JK Flipflop when $\mathrm{J}=\mathrm{K}=1$.
This may lead to unpredictable output when clock disables and it is called as race around condition.
This can be avoided by

1. Using edge triggered JK flipflop
2. Using Master Slave JK flipflop

In edge triggered JK flipflop, race condition is avoided as the enable input of the clock remains active only for a short time.

In master slave flipflop master and slave are inverted clock so that only one will be active at a time. During the high of the clock, output will be toggled and will remain at the output of master but could not be followed by slave as it is disabled. So, output of slave remains same as earlier and toggled output remains at master. When clock is low master is deactivated and slave is enabled which only follows the output of master. So at the end of one complete clock we just have one toggled output and race around condition is avoided.

Q4] a) Design 3 bit asynchronous counter and draw the timing diagram.

## Solution :-

A ripple counter is an asynchronous counter where only the first flip-flop is clocked by an external clock. All subsequent (next occurring) flip-flops are clocked by the output of the preceding flip-flop.

In asynchronous counter, a clock pulse drives FFO. Output of FFO drives FF1 which then drives the FF2 flip flop. All J and K inputs are connected to Logic 1. Therefore, each flip flop will toggle with negative transition at its clock input.


Fig. 4.1 bit Asynchronous counter
The 3 bit MOD-8 asynchronous counter consists of 3 JK flip flops. Overall propagation delay time is the sum of individual delays. Initially all flip flops are reset to produce 0 . The output conditions is Q2 $\mathrm{Q} 1 \mathrm{Q} 0=000$.

When the first clock pulse is applied, the FFO changes state on its negative edge. Therefore, Q2Q1Q0 $=001$. On the negative edge of second clock pulse flip flop FFO toggles. Its output changes from 1 to 0 . This being negative change, FF1 changes state. Therefore, Q2 Q1 Q0 $=010$. Similarly, the output of flipflop FF2 changes only when there is negative transition at its input when fourth clock pulse is applied.


Fig. 4.2 Timing diagram
The output of the flip flops is a binary number equivalent to the number of clock pulses received. The output conditions are as shown in the truth table.

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| Counter <br> State | Q2 | Q1 | Qo |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 1 |
| 2 | 0 | 1 | 0 |
| 3 | 1 | 1 | 1 |
| 4 | 1 | 0 | 0 |
| 5 | 1 | 1 | 1 |
| 7 |  | 0 | 1 |
| 7 |  |  |  |

On the negative edge of eighth pulse, counter is reset. The counter acts as a frequency divider. FFO divides clock frequency by 2, FF1 divides clock frequency by 4 , FF2 divides clock frequency by 8 . If $n$ flip flops are cascaded, we get 2 n output conditions. The largest binary number counted by n cascaded flip flops has a decimal equivalent of $2 n-1$. MOD-8 counter has count of the largest binary number 111 which has decimal equivalent of 23-1=7.

Q4] b) Convert JK flipflop to SR flipflop and D flipflop.

## Solution :-

JK to SR
a) Excitation Table

| Input's |  | Present State | Next State | Flip Flop Input |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S | R | Qn | Qn+1 | J | K |
| 0 | 0 | 0 | 0 | 0 | X |
| 0 | 0 | 1 | 1 | X | 0 |
| 0 | 1 | 0 | 0 | X | X |
| 0 | 1 | 1 | 0 | X | 1 |
| 1 | 0 | 0 | 1 | 1 | X |
| 1 | 0 | 1 | 1 | X | 0 |
| 1 | 1 | 0 | X | X | X |
| 1 | 1 | 1 | X | X | X |

b) K- map simplification

For J

| Qn SR | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | $x$ | 1 |
| 1 | $x$ | $x$ | $x$ | $x$ |

For K

| Qn SR | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 0 | $\times$ | $\times$ | $\times$ | 0 |
| 1 | 0 | 1 | $\times$ | $\times$ |

c) Logic diagram


## JK to D

a) Excitation Table

| Input | Present State | Next State | Flip Flop Input |  |
| :---: | :---: | :---: | :---: | :---: |
| D | Qn | Qn+1 | J | K |
| 0 | 0 | 0 | 0 | X |
| 0 | 1 | 0 | X | 1 |
| 1 | 0 | 1 | 1 | X |
| 1 | 1 | 1 | X | 0 |

b) K-map simplification

For J

| $D$ | 0 | 1 |
| :---: | :---: | :---: |
| $Q_{n}$ | 0 | 1 |
| 0 | $\times$ | $\times$ |
| 1 | $\mathrm{~J}=\mathrm{D}$ |  |

For K

| $D$ | 0 | 1 |
| :---: | :---: | :---: |
| 0 | $\left[\begin{array}{c}X \\ 0\end{array}\right.$ | $\times$ |
| 1 | 1 | 0 |

c) Logic diagram


Q5] a) Compare TTL and CMOS with respect to different parameters.

## Solution :-

| PARAMETER | TTL | CMOS |
| :---: | :---: | :---: |
| Basic gate | NAND | INVERTER |
| Circuit | Transistor- transistor | Complementary MOS |
| Speed of operation | More | Less |
| Propagation delay per rate | 4-12 ns | 50 ns |
| Power supply voltage | $5 \mathrm{~V}$ | 3 to 15V |
| Power dissipation | $\mathrm{P}_{\mathrm{D}}=10 \mathrm{Mw}$ | $\mathrm{P}_{\mathrm{D}}=0.1 \mathrm{~mW}$ |
| Wired collector capability | With passive pull up | With tristate output |
| Fan-out | 10 | 50 |
| Noise Immunity | Good | Very good to excellent |
| Speed power product | 100 pJ | 10.5 pJ |
| Available functions | Very high | High |
| Compatibility with other families | With DTL | No, but compatible with TTL for 5V supply |
| Switching speed | Faster than CMOS | Less than TTL |
| Circuits | It uses BJT's | It uses FET's |
| Noise immunity | Less than CMOS | Better than TTL |
| Power consumption | It consumes lot more power than CMOS chips. | Consumes less power than TTL chips. |

## Q5 b) Explain the features of VHDL and its modeling styles.

## Solution:-

VHDL stands for very high-speed integrated circuit hardware description language. This language was first introduced in 1981 for the department of Defense (DoD) under the VHSIC program. Features of VHDL are as follows:

1) Designs are organized hierarchically.
2) Each design element has:
i) A well-defined interface.
ii) A precise behavioral specification using either algorithmic description or hardware structural description.
3) Models concurrency, timing, and clocking:
i) Handles asynchronous and synchronous circuits
ii) Designs can be simulated.
4) Language is not case sensitive.
5) A formal language for specifying the behavior and structure digital circuit.

Modeling styles: Modeling Style means, that how we Design our Digital IC's in Electronics. With the help of modeling style we describe the Design of our Electronics.

## Normally we use Three type of Modeling Style in VHDL

- Data Flow Modeling Style.
- Behavior Modeling Style.
- Structural Modeling Style.

Data Flow Modeling Style - Dataflow system describes a system in terms of how data flows through the system. Data dependencies in the description match those in atypical hardware implementation. A dataflow description directly implies a corresponding gate level implementation.
Dataflow descriptions consist of one or more concurrent signal assignment situation
Behavior Modeling Style - A behavioral description describes a system's behavior or function in an algorithmic fashion.
Behavioral style is the most abstract style. The description is abstract in the sense that it does not directly imply a particular gate-level implementation.
Behavioral style consists of one or more process statements. Each process statement is a single concurrent statement that itself contains one or more sequential statements.
Sequential statements are executed sequentially statements by a simulator, the same as the execution of sequential statements in a conventional programming language.
Structural Modeling Style - In structural style of modelling, an entity is described as a set of interconnected components.
The top-level design entity's architecture describes the interconnection of lower level design entities. Each lower level entity can, in turn, be described as an interconnection of design entities at the nextlower level, and so on.

Structural style is most useful and efficient when a complex system is described as an interconnection of moderately complex design entities. This approach allows each design entity to be independently designed and verified before being used in the higher-level description.

## Q6] a) Moore and Mealy machine.

## Solution :-

Moore Machines: Moore machines are finite state machines with output value and its output depends only on present state. It can be defined as $(Q, q 0, \Sigma, 0, \delta, \lambda)$ where:

- Q is finite set of states.
- q 0 is the initial state.
- $\quad \Sigma$ is the input alphabet.
- O is the output alphabet.
- $\delta$ is transition function which maps $\mathrm{Q} \times \Sigma \rightarrow \mathrm{Q}$.
- $\lambda$ is the output function which maps $Q \rightarrow 0$.


Figure 1
In the moore machine shown in Figure 1, the output is represented with each input state separated by /. The length of output for a moore machine is greater than input by 1.

- Input: 11
- Transition: $\delta(q 0,11)=>\delta(q 2,1)=>q 2$
- Output: 000 ( 0 for q0, 0 for q2 and again 0 for q2)

Mealy Machines: Mealy machines are also finite state machines with output value and its output depends on present state and current input symbol. It can be defined as ( $Q, q 0, \Sigma, 0, \delta, \lambda^{\prime}$ ) where:

- $\quad Q$ is finite set of states.
- q0 is the initial state.
- $\Sigma$ is the input alphabet.
- O is the output alphabet.
- $\delta$ is transition function which maps $\mathrm{Q} \times \Sigma \rightarrow \mathrm{Q}$.
- ' $\lambda$ ' is the output function which maps $\mathrm{Q} \times \Sigma \rightarrow 0$.

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Figure 2
In the mealy machine shown in Figure 1, the output is represented with each input symbol for each state separated by /. The length of output for a mealy machine is equal to the length of input.

- Input:11
- Transition: $\delta(q 0,11)=>\delta(q 2,1)=>q 2$
- Output: 00 ( $q 0$ to $q 2$ transition has Output 0 and $q 2$ to $q 2$ transition also has Output 0 )


## Q6] b) Sequence generator.

## Solution :-

the sequence generators are nothing but a set of digital circuits which are designed to result in a specific bit sequence at their output. There are several ways in which these circuits can be designed including those which are based on multiplexers and flip-flops.

A circuit which generates a prescribed sequence of bits, in synchronization with a clock, is referred to as a sequence generator such generator can be used as:

1] counters.
2] Random bit generators.
3] Prescribed period and sequence generators.
4] Code generator.

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The basic structure of a sequence generator is shown in diagram below:


Basic Structure of a Sequence Generator.
The output Y of the next state decoder is function of $\mathrm{QN}-2, \mathrm{QN}-1$ $\qquad$ Q1, Q0. This system is similar to a ring counter or twisted ring counter which are special case of sequence generator, the design of the decoder will be clear.

## Q6] c) Universal shift register.

## Solution:-

A Universal shift register is a register which has both the right shift and left shift with parallel load capabilities. Universal shift registers are used as memory elements in computers. A Unidirectional shift register is capable of shifting in only one direction. A bidirectional shift register is capable of shifting in both the directions. The Universal shift register is a combination design of bidirectional shift register and a unidirectional shift register with parallel load provision. n-bit universal shift register -
A $n$-bit universal shift register consists of $n$ flip-flops and $n 4 \times 1$ multiplexers. All the $n$ multiplexers share the same select lines(S1 and S0)to select the mode in which the shift register operates. The select inputs select the suitable input for the flip-flops.
The working of the Universal shift register depends on the inputs given to the select lines.
The register operations performed for the various inputs of select lines are as follows:

| S1 | SO | REGISTER OPERATION |
| :---: | :---: | :---: |
| 0 | 0 | No changes |
| 0 | 1 | Shift Right |
| 1 | 0 | Shift Left |

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| 1 | 1 | Parallel Load |
| :---: | :---: | :---: |

## Q6] d) Priority Encoder.

## Solution :-

The priority encoders output corresponds to the currently active input which has the highest priority. So, when an input with a higher priority is present, all other inputs with a lower priority will be ignored.

A priority encoder provides $n$ bits of binary coded output representing the position of the highest order active input of $2^{\text {n }}$ inputs. If two or more inputs are high at the same time, the input having the highest priority will take precedence.


Fig.6.d Priority Encoder
A 4-to-2 priority encoder takes 4 input bits and produces 2 output bits as shown in figure 6.4. In this truth table, for all the non-explicitly defined input combinations (i.e. inputs containing 2,3 , or 4 high bits) the lower priority bits are shown as don't cares (X). Similarly, when the inputs are 0000, the outputs are not valid and therefore they are XX.

Truth table for 4 bit input to priority encoder:

| I3 | I2 | I1 | IO | Y1 | YO |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | $X$ | $X$ |
| 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | $X$ | $X$ | 1 |
| 0 | 1 | $X$ | $X$ | 1 | 0 |
| 0 | $X$ | $X$ |  |  |  |

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| 1 | $X$ | $X$ | $X$ | 1 | 1 |
| :--- | :--- | :--- | :--- | :--- | :--- |

From this truth table, we use the Karnaugh Map to minimize the logic to the following Boolean expressions:
$Y 1=12+13$
$Y O=13+12 \times 11$

## Q6] e) Carry look ahead adder

## Solution :-

In ripple carry adders, for each adder block, the two bits that are to be added are available instantly. However, each adder block waits for the carry to arrive from its previous block. So, it is not possible to generate the sum and carry of any block until the input carry is known. This disadvantage is overcomed in carry look ahead adder.

Carry Look-ahead Adder:
A carry look-ahead adder reduces the propagation delay by introducing more complex hardware. In this design, the ripple carry design is suitably transformed such that the carry logic over fixed groups of bits of the adder is reduced to two-level logic.


Fig 6.e Carry look ahead adder
We define two variables as 'carry generate' Gi and 'carry propagate' Pi then,
$\mathrm{Pi}=\mathrm{Ai} \oplus \mathrm{Bi}$
$\mathrm{Gi}=\mathrm{Ai} \times \mathrm{Bi}$

